

No. 3679

LA4583M

Compact Cassette Stereo Recording and Playback System

OVERVIEW

The LA4583M is a complete recording and playback system for compact cassette units. It incorporates microphone and tuner preamplifiers, and power amplifiers for both a low-impedance speaker and headphones. Mode switching is completely electronic, which greatly reduces switching noise.

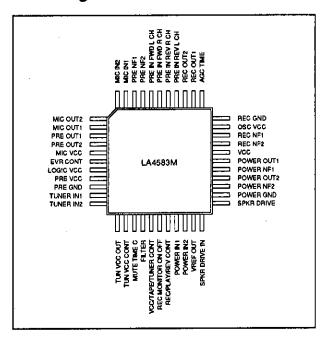
The LA4583M features ALC circuitry to ensure optimum recording levels and on-chip supplies for external electret condenser microphones and bias oscillators.

The LA4583M operates from a 3 V supply and is available in 44-pin QIPs.

FEATURES

- Electronically switched recording and playback modes
- Automatic power and recording mute when switching modes
- 16 Ω stereo headphone amplifier
- 4 Ω mono speaker amplifier
- · Automatic level control in recording mode
- Preamplifiers for external microphones and tuner circuits
- 3 V supply
- 44-pin QIP

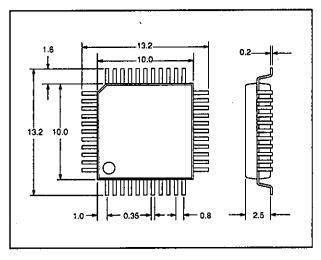
Pin Assignment



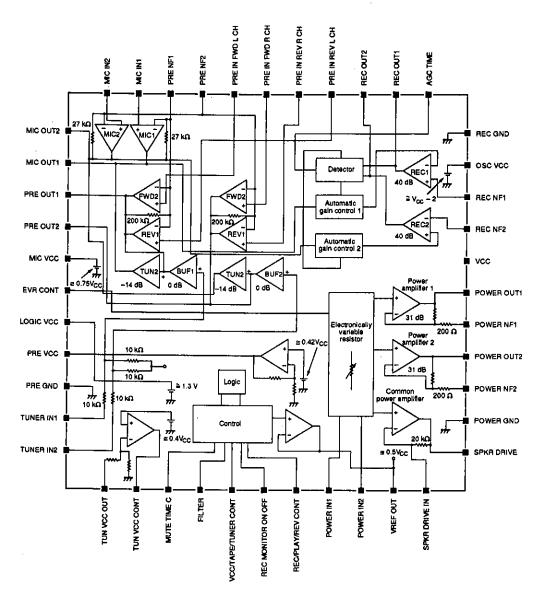
PACKAGE DIMENSIONS

Unit: mm

3148-QIP44MA



BLOCK DIAGRAM



PIN DESCRIPTION

Number	Name	Equivalent circuit	Description
1	TUN VCC OUT	10 kg	Tuner supply output. Output impedance is 20 k Ω . Nominal voltage is 2.5 V.
2	TUN VCC CONT	50 kg VCC	Tuner supply control output. Output impedance is 51 k Ω . Nominal voltage is 2.4 V.

Number	Name	Equivalent circuit	Description
3	MUTE TIME C	10 KD 10 KD GND	Mute time control input. Nominal voltage range is 0.2 to 0.85 V.
4	FILTER	10 kΩ TR1 10 kΩ 10 kΩ 17 kΩ 17 kΩ 17 kΩ 17 kΩ 17 kΩ 17 kΩ 18 kΩ 17 kΩ	Voltage reference external filter capacitor. Nominal voltage is 1.2 V.
5	VCC/TAPE/TUNER CONT	20 kg 25 kg GND	Tape/tuner mode select input. Nominal voltage is 1.1 to 3.0 V.
6	REC MONITOR ON OFF	200 Ω 1 kΩ GND	Recording monitor enable input. Input impedance is 110 kΩ. Nominal voltage is 0 to 0.7 V.
7	REC/PLAY/REV CONT	VCC VREF OUT 20 kΩ 50 kΩ 20 kΩ 20 kΩ 50 kΩ 60 kΩ	Recording/playback/reverse-playback mode select. Input impedance is 75 kΩ. Nominal voltage range is 0.1 to 1.5 V.
8	POWER IN1	25 kg 10 kg	Channel 1 headphone amplifier input. Input impedance is 50 kΩ. Nominal voltage is 1.5 V.
9	POWER IN2	Power IN1/2 VREF	Channel 2 headphone amplifier input. Input impedance is 50 kΩ. Nominal voltage is 1.5 V.

Number	Name	Equivalent circuit	Description		
10	VREF OUT	VCC GND	Voltage reference output. Nominal voltage is 1.5 V.		
11	SPKR DRIVE IN	200 Ω VREF SPKR DRIVE GND	Common amplifier input. Input impedance is 5 kΩ. Nominal voltage is 1.5 V.		
12	SPKR DRIVE	VCC VCC	Common amplifier output. Nominal voltage is 1.5 V.		
15	POWER OUT2	20 kΩ	Channel 2 headphone amplifier output. Nominal voltage is 1.5 V.		
17	POWER OUT1	GND	Channel 1 headphone amplifier output. Nominal voltage is 1.5 V.		
13	POWER GND		Power amplifier ground		
14	POWER NF2	200 a	Channel 2 headphone amplifier negative feedback input. Nominal voltage is 1.5 V.		
16	POWER NF1	From electronically o variable resistor OUT GND	Channel 1 headphone amplifier negative feedback input. Nominal voltage is 1.5 V.		
18	VCC		5 V supply		
19	REC NF2	VCC VCC	Channel 2 recording preamplifier negative feedback input. Nominal voltage is 1.5 V.		
20	REC NF1	From audio level o control	Channel 1 recording preamplifier negative feedback input. Nominal voltage is 1.5 V.		

Number	Name	Equivalent circuit	Description
21	osc vcc	10 KΩ VCC	External bias oscillator supply. Output impedance is 20 kΩ. Nominal voltage is 0.8 V.
22	REC GND		Recording amplifier ground
23	AGC TIME	O VCC To automatic o gain control	AGC response control input. Nominal voltage range is 0.35 to 0.75 V.
24	REC OUT1	50 NO Detector	Channel 1 recording amplifier output. Nominal voltage is 1.5 V.
25	REC OUT2	SO KO VREF VREF	Channel 2 recording amplifier output. Nominal voltage is 1.5 V.
26	PRE IN REV LCH	vcc vcc	Left-channel reverse playback preamplifier input. Nominal voltage is 1.5 V.
27	PRE IN REV RCH		Right-channel reverse playback preamplifier input. Nominal voltage is 1.5 V.
28	PRE IN FWD RCH	200 Ω To feedback	Left-channel normal playback preamplifier input. Nominal voltage is 1.5 V.
29	PRE IN FWD LCH	resistor	Right-channel normal playback preamplifier input. Nominal voltage is 1.5 V.
30	PRE NF2	200,0	Channel 2 preamplifier negative feedback input. Nominal voltage is 1.5 V.
31	PRE NF1	PRE OUT MIC OUT GND	Channel 1 preamplifier negative feedback input. Nominal voltage is 1.5 V.

Number	Name	Equivalent circuit	Description
32	MIC IN1	PRE VCC	Channel 1 microphone preamplifier input. Input impedance is 5 k Ω . Nominal voltage is 1.5 V.
33	MIC IN2	VREF GND	Channel 2 microphone preamplifier input. Input impedance is 5 k Ω . Nominal voltage is 1.5 V.
34	MIC OUT2	PRE VCC PRE OUT 200 kΩ PRE NF 27 kΩ	Channel 2 microphone preamplifier output. Nominal voltage is 1.5 V.
35	MIC OUT1	gnd	Channel 1 microphone preamplifier output. Nominal voltage is 1.5 V.
36	PRE OUT1	PRE NF PRE VCC 200 κΩ 200 Ω	Channel 1 playback preamplifier output. Nominal voltage is 1.5 V.
37	PRE OUT2	GND GND	Channel 2 playback preamplifier output. Nominal voltage is 1.5 V.
38	MIC VCC	ονςς 6 κΩ 11 κΩ • GND	External microphone supply. Output impedance is 18 kΩ. Nominal voltage is 2.3 V.
39	EVR CONT	LOGIC VCC 10 kΩ 8.2 kΩ 15 kΩ eGNO	Electronic variable resistor (EVR) control voltage input. Input impedance is 16 kΩ. Nominal voltage is 0 to 1.3 V.
40	LOGIC VCC	EVR CONT	Bias supply for internal logic circuits. Nominal voltage is 1.3 V.

Number	Name	Equivalent circuit	Description
41	PRE VCC	VCC 4.5 KΩ FILTER GND	Bias supply for preamplifiers. Output impedance is 9.5 kΩ. Nominal voltage is 2.5 V.
42	PRE GND		Ground
43	TUNER IN1	VCC PRE OUT	Channel 1 tuner preamplifier input. Input impedance is 20 kΩ. Nominal voltage is 1.5 V.
44	TUNER IN2	10 kΩ 33 kΩ VREF OUT GND	Channel 2 tuner preamplifier input. Input impedance is 20 kΩ. Nominal voltage is 1.5 V.

Note

Nominal voltages measured with V_{CC} = 3 V.

SPECIFICATIONS

Absolute Maximum Ratings

Parameter	Symbol	Rating	Unit
Supply voltage	Vœ	4.5	٧
Power dissipation	P _D	1	W
Operating temperature range	Topr	-20 to 75	°C
Storage temperature range	T _{stg}	–40 to 125	°C

Recommended Operating Conditions

 $T_n = 25^{\circ} C$

Parameter	Symbol	Rating	Unit
Supply voltage	Vœ	3	٧
Supply voltage range	Vœ	1.8 to 3.6	٧

Electrical Characteristics

 V_{cc} = 3 V, T_a = 25 °C, 0 dB = 1 V, f = 1 kHz, R_L = 10 k Ω (preamplifiers), R_L = 16 Ω (headphone amplifiers), R_L = 4 Ω (common amplifier)

Parameter	Symbol	Condition		Unit		
		Containon	min	typ	max	Olin
Quiescent supply current	logo	Forward playback mode	14	22	40	A
Current	Icco	Recording mode	14	28	45	- mA

Playback preamplifiers

Parameter	Symbol	Condition	Rating			,,,,
			min	typ	max	Unit
Open-loop voltage gain	VG _{HO}		65	80	_	dB
Closed-loop voltage gain	VG _H		36	38	40	dB
Total harmonic distortion	THD _H	V ₀ = -5 dBV	-	0.05	1.0	%
Input conversion noise	V _{NIH}	$R_0 = 2.2 \text{ k}\Omega$, DIN AUDIO filter	-	1	2	μV
Maximum output voltage	V _{OH} max	THD = 5%	0.3	0.65	_	v

Tuner preamplifiers

Parameter	Symbol	Condition	Rating			1715
T WI GITISTO	- Jimaoi	Condition	min	typ	max	Unit
Closed-loop voltage gain	VGR		-8	6	-4	dB
Total harmonic distortion	THD _R	$V_0 = -15 \text{ dBV}$	-	0.1	1.0	%
Maximum output voltage	V _{OR} max	THD = 5%	0.2	0.6	-	V
Output noise voltage	V _{NOR}	$R_0 = 10 \text{ k}\Omega$, DIN AUDIO filter	-	10	30	μ۷

Microphone recording circuitry

Parameter	Symbol	Condition		11-14		
Total harmonic distortion Maximum output voltage			min	typ	max	Unit
Closed-loop voltage gain	VG _{MR}		77	80	83	dB
Total harmonic distortion	THD _{MR}	V ₀ = 0.775 V, C35 = C36 = 1.8 nF	-	0.7	1.5	%
Maximum output voltage	V _{OMR}	THD = 5%, AGC is OFF.	0.7	1.1	_	V
Output noise voltage	VNOMR	$R_0 = 2.2 \text{ k}\Omega$, DIN AUDIO filter	-	15	30	m∨
Channel balance	CB _{MR}	,	-2	0	2	dB
Interchannel crosstalk	CT _{MR}	$R_0 = 2.2 \text{ k}\Omega$, 1 kHz TUNE filter	16	25	-	dB

Tuner recording circuitry

Parameter	Symbol Condition	Condition	Rating			
		Condition	min	typ	max	Unit
Closed-loop voltage gain	VGRR	-10 dBV at tuner inputs	19	21	23	dB
Total harmonic distortion	THD _{RR}	V _O = -6 dBV, EVR is OFF.	_	0.15	1.0	%
Output voltage	V _{ORR} max	THD = 5%	0.7	1.1	-	٧
Channel balance	CB _{RR}		-2	0	2	dB
Interchannel crosstalk	CTAR	R _g = 10 kΩ, 1 kHz TUNE filter	30	50	-	dB

EVR and power amplifiers

Parameter	Symbol Condition	Condition	Rating			*1-14
		Constituti	min	typ	max	Unit
Closed-loop voltage gain	VGE	EVR is at maximum resistance.	28	31	34	dB
Total harmonic distortion	THDE	P ₀ = 10 mW	0.1	0.5	1.0	%
Maximum output power	Po max	THD = 10%	20	33	-	mW
Output noise voltage	Vnore	EVR is OFF. DIN AUDIO filter	-	170	500	μV.
Maximum EVR attenuation	CLATT	EVR is OFF. 1 kHz TUNE filter	50	70	_	dB
EVR input impedance	Riev		35	50	65	kΩ

Preampilfiers and power amplifiers

Parameter	Symbol Condition	Condition	Rating			1
		min	typ	max	Unit	
Channel balance	CB _⊤		-2	0	2	dB
Interchannel crosstalk	CTT	$R_g = 2.2 \text{ k}\Omega$, 1 kHz TUNE filter	30	38	_	dB
Crosstalk between forward and reverse playback preamplifiers	СТнн	1 kHz TUNE filter	30	65	-	dB
Crosstalk between playback and tuner inputs	CTHR	1 kHz TUNE filter	50	70	_	dB
Crosstalk between recording amplifier outputs and microphone inputs	CT _{MRH}	Recording mode, 1 kHz TUNE filter	50	70	_	dB

Automatic gain control

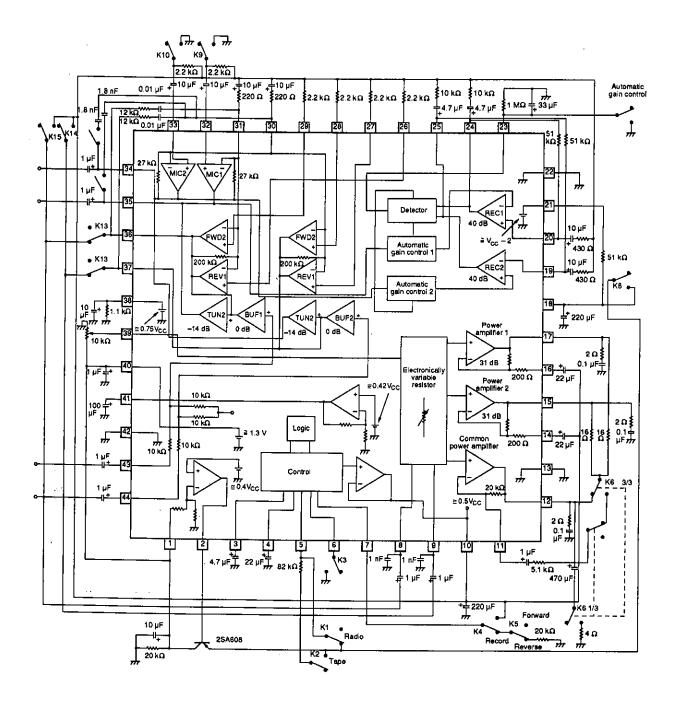
Parameter	Symbol	Condition	Rating			I I I I I I I I I I I I I I I I I I I
			min	typ	max	Unit
AGC response	AGCW	From activation to 5% THD at RECOUT	35	42	_	dB
Channel balance	AGC _{BLC}		-2	0	2	dB
Minimum activation level	AGCINS		-30	-27	-24	dBV

Common amplifier

Parameter	Symbol	Condition	Rating			
		Condition	min	typ	max	Unit
Closed-loop voltage gain	VG _{PWA}	$R_{XR} = 5 k\Omega$	9	11	13	dB
Total harmonic distortion	THD _{PWR}	P ₀ = 50 mW	-	0.6	1.0	%
Output noise voltage	VNOPWR	$R_{XR} = 5 k\Omega$, DIN AUDIO filter	_	15	50	μ۷
Maximum output power	Popwr	THD = 10%	75	120	_	mW
Microphone supply voltage	V _{MCC}	IL = 2 mA	2.0	2.3	2.5	V

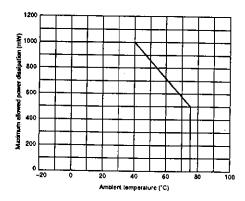
Parameter	Symbol	Condition	Rating			11-14
	Oymbol		min	typ	max	Unit
Tuner supply voltage	V _{TCC}	I _L = 30 mA, external transistor is 2SA1178.	2.0	2.5	2.7	٧
Oscillator supply voltage	Voscoc	I _L = 15 mA	0.65	0.85	1.0	V

Measurement Circuit

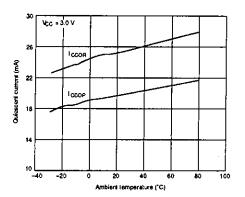


Typical Performance Characteristics

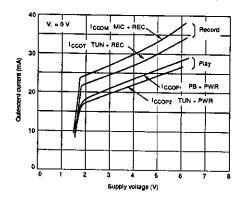
Power dissipation vs. ambient temperature



Quiescent current vs. ambient temperature

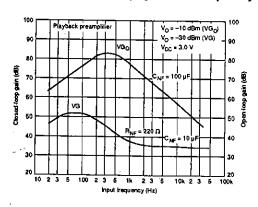


Quiescent current vs. supply voltage

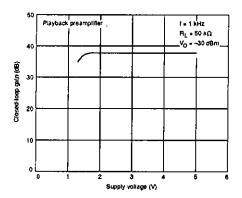


Playback preamplifiers

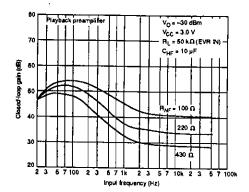
Closed-loop and open-loop gain vs. frequency



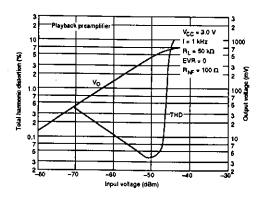
Closed-loop gain vs. supply voltage



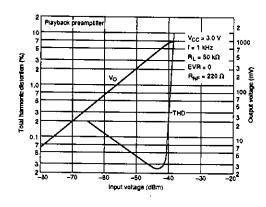
Closed-loop gain vs. frequency



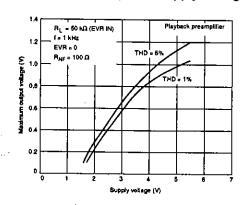
Total harmonic distortion and output voltage vs. input voltage (1)



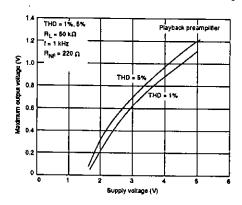
Total harmonic distortion and output voltage vs. Input voltage (2)



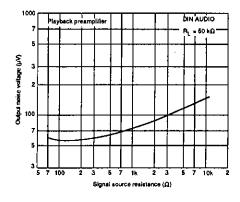
Maximum output voltage vs. supply voltage (1)



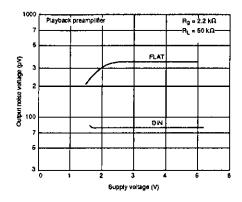
Maximum output voltage vs. supply voltage (2)



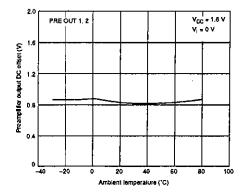
Output noise voltage vs. source resistance



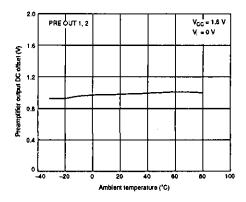
Output noise voltage vs. supply voltage



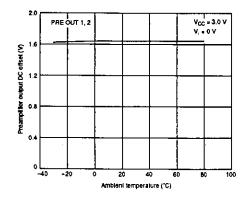
Output DC offset vs. ambient temperature (1)



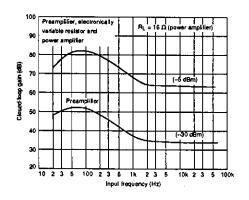
Output DC offset vs. ambient temperature (2)



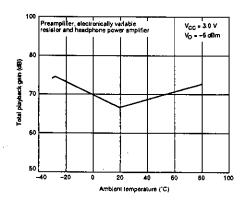
Output DC offset vs. ambient temperature (3)



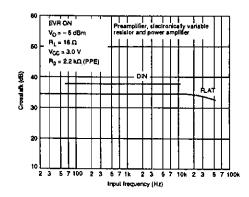
Preamplifiers, EVR and power amplifiers Closed-loop gain vs. frequency



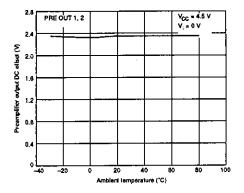
Total playback gain vs. ambient temperature



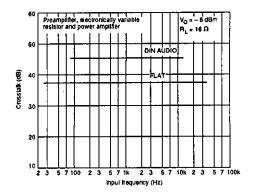
Crosstalk between channels vs. frequency



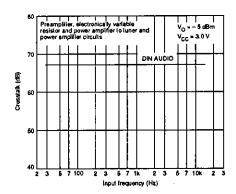
Output DC offset vs. ambient temperature (4)



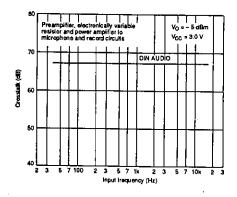
Crosstalk between forward and reverse channels vs. frequency



Crosstalk with tuner inputs vs. frequency

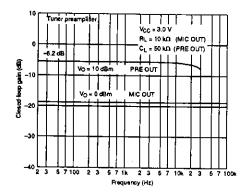


Crosstalk with microphone and recording amplifiers vs. frequency



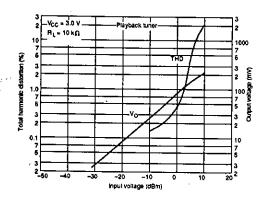
Tuner preamplifiers

Closed-loop gain vs. frequency

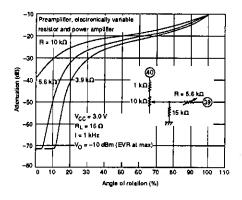


Playback and tuner preamplifiers

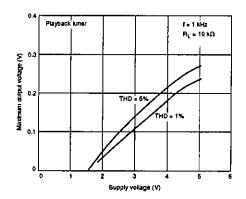
Total harmonic distortion and output voltage vs. input voltage



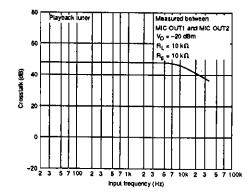
EVR attenuation vs. potentiometer rotation



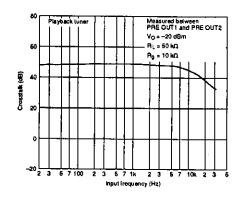
Maximum output voltage vs. supply voltage



Crosstalk between channels vs. frequency (1)

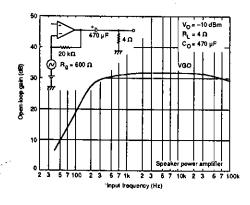


Crosstalk between channels vs. frequency (2)

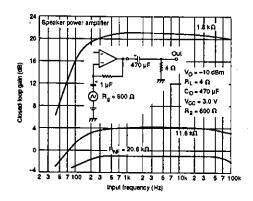


Common amplifier

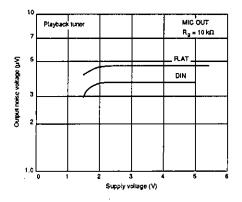
Open-loop voltage gain vs. frequency



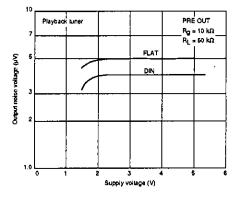
Closed-loop voltage gain vs. frequency (1)



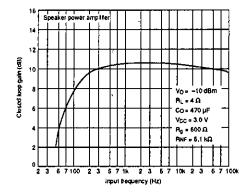
Output noise voltage vs. supply voltage (1)



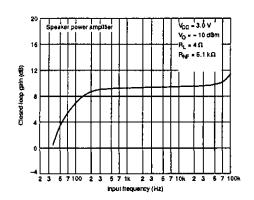
Output noise voltage vs. supply voltage (2)



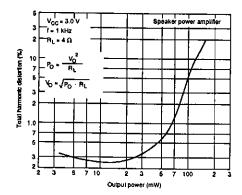
Closed-loop voltage gain vs. frequency (2)



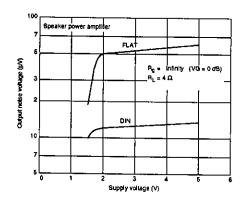
Closed-loop voltage gain vs. frequency (3)



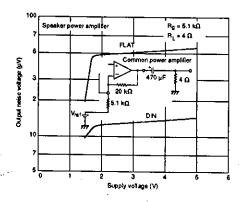
Total harmonic distortion vs. output power



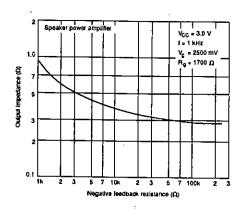
Output noise voltage vs. supply voltage (1)



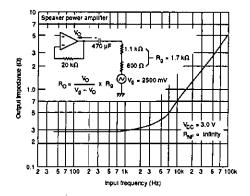
Output noise voltage vs. supply voltage (2)



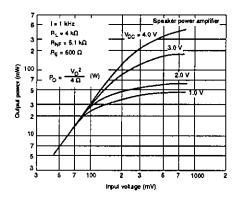
Output impedance vs. negative feedback resistance



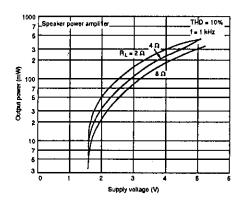
Output Impedance vs. frequency



Output power vs. input voltage

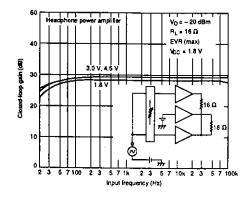


Maximum output power vs. supply voltage

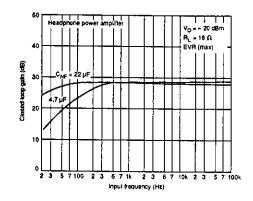


Headphone amplifiers

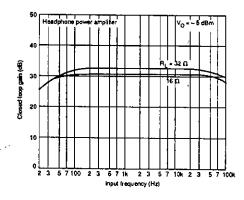
Closed-loop voltage gain vs. frequency (1)



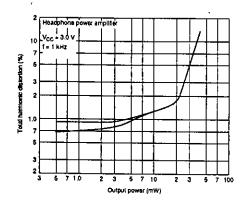
Closed-loop voltage gain vs. frequency (2)



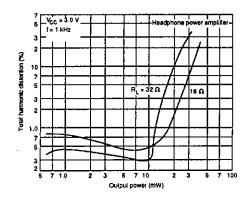
Closed-loop voltage gain vs. frequency (3)



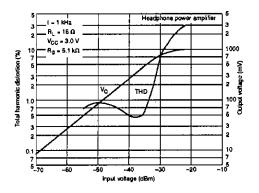
Total harmonic distortion vs. output power (1)



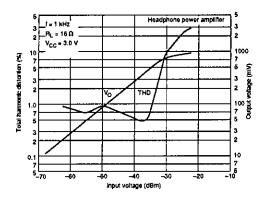
Total harmonic distortion vs. output power (2)



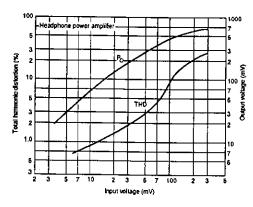
Total harmonic distortion and output voltage vs. input voltage (1)



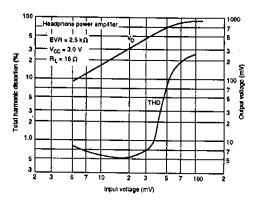
Total harmonic distortion and output voltage vs. input voltage (2)



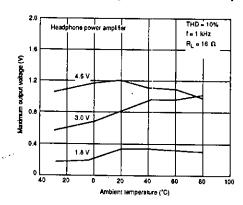
Total harmonic distortion and output voltage vs. input voltage (3)



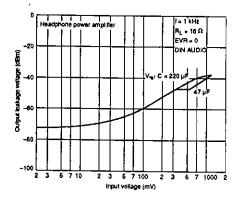
Total harmonic distortion and output voltage vs. input voltage (4)



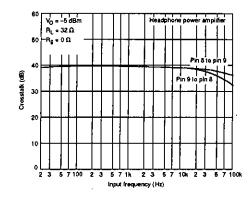
Maximum output voltage vs. ambient temperature



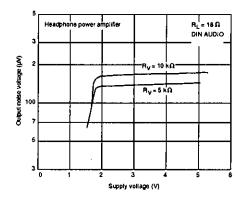
Output voltage vs. input voltage



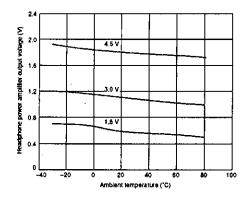
Crosstalk between channels vs. frequency



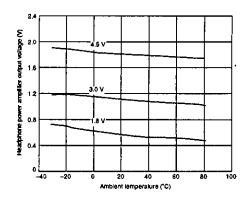
Output noise voltage vs. supply voltage



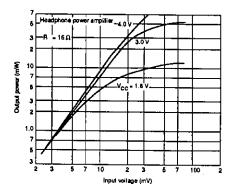
Output voltage vs. ambient temperature (1)



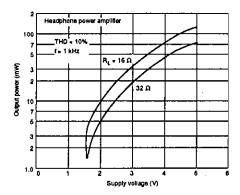
Output voltage vs. ambient temperature (2)



Output power vs. Input voltage

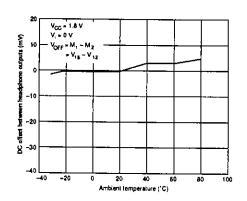


Output power vs. supply voltage

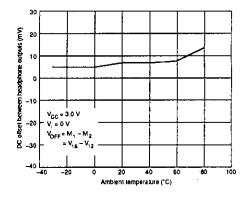


Headphone and common amplifier

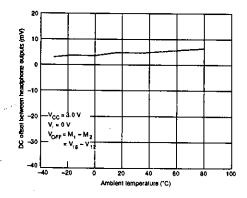
DC offset between headphone outputs vs. ambient temperature (1)



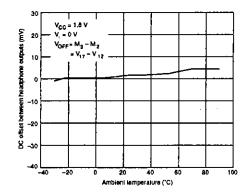
DC offset between headphone outputs vs. ambient temperature (3)



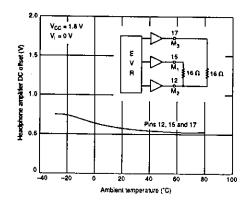
DC offset between headphone outputs vs. ambient temperature (2)



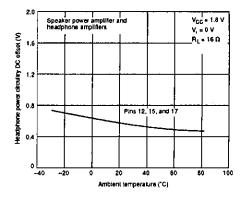
DC offset between headphone outputs vs. ambient temperature (4)



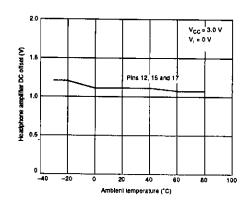
Amplifier DC offset vs. ambient temperature (1)



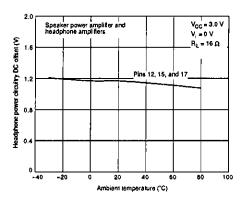
Amplifler DC offset vs. ambient temperature (5)



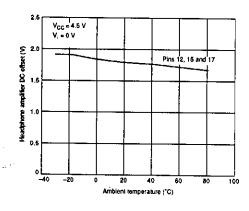
Amplifier DC offset vs. ambient temperature (2)



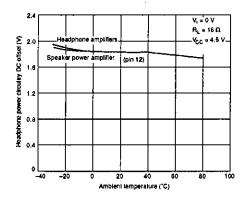
Amplifier DC offset vs. ambient temperature (6)



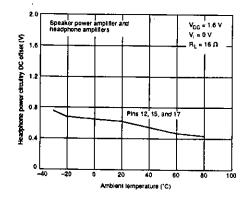
Amplifier DC offset vs. ambient temperature (3)



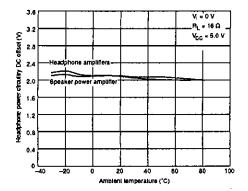
Amplifier DC offset vs. ambient temperature (7)



Amplifler DC offset vs. ambient temperature (4)

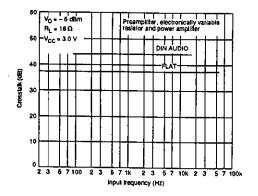


Amplifier DC offset vs. ambient temperature (8)



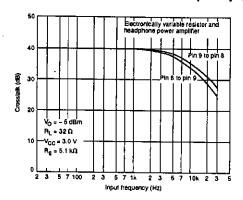
Playback amplifiers, EVR and power amplifiers

Crosstalk between forward and reverse channels vs. frequency



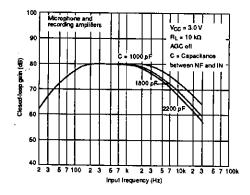
EVR and headphone amplifiers

Crosstalk between channels vs. frequency (1)

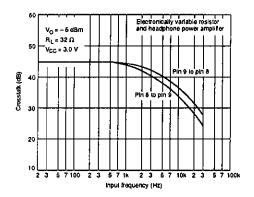


Microphone and recording amplifiers

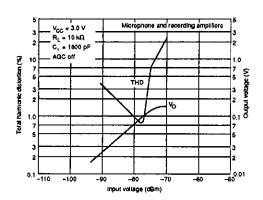
Closed-loop gain vs. frequency



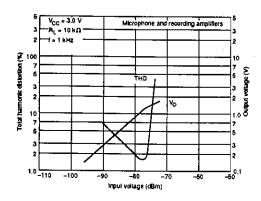
Crosstalk between channels vs. frequency (2)



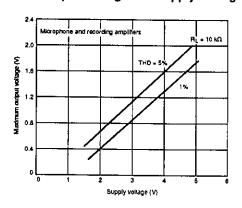
Total harmonic distortion and output voltage vs. input voltage (1)



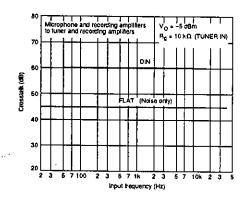
Total harmonic distortion and output voltage vs. input voltage (2)



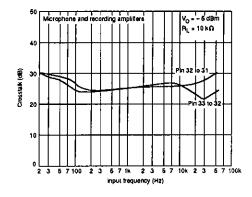
Maximum output voltage vs. supply voltage



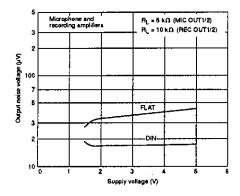
Crosstalk with tuner and recording ampliflers vs. frequency



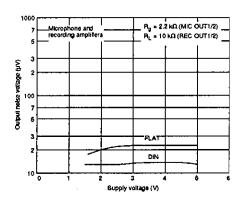
Crosstalk between channels vs. frequency



Output noise voltage vs. supply voltage (1)

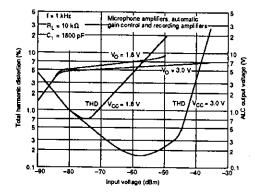


Output noise voltage vs. supply voltage (2)

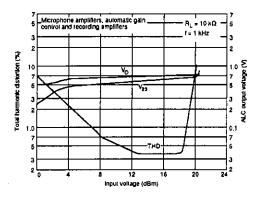


Microphone amplifiers, AGC and recording amplifiers

Total harmonic distortion and ALC output voltage vs. input voltage (1)

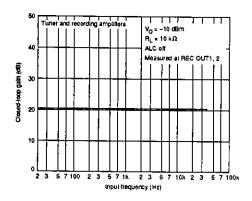


Total harmonic distortion and ALC output voltage vs. Input voltage (2)

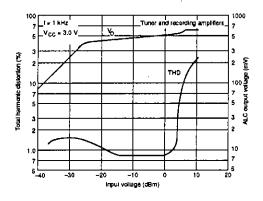


Tuner and recording amplifiers

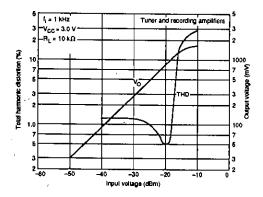
Closed-loop voltage gain vs. frequency



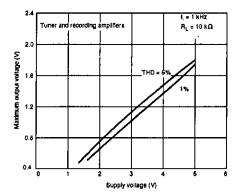
Total harmonic distortion and ALC output voltage vs. Input voltage



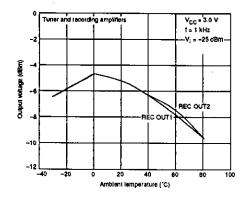
Total harmonic distortion and output voltage vs. input voltage



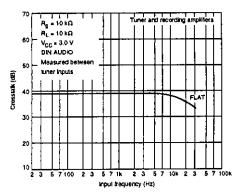
Maximum output voltage vs. supply voltage



Output voltage vs. ambient temperature

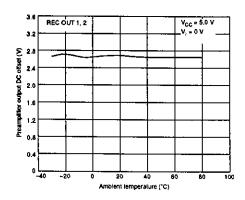


Crosstalk between channels vs. frequency



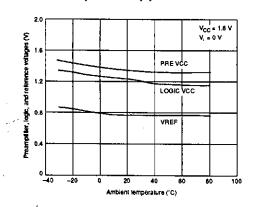
Recording amplifier

DC offset vs. ambient temperature

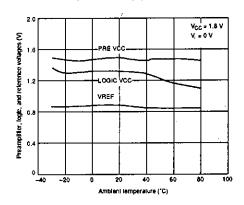


Reference voltages

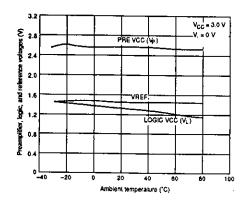
 V_{REF} , preamplifier and logic reference voltages vs. ambient temperature (1)



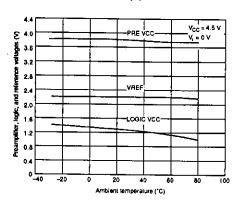
 V_{REF} , preamplifier and logic reference voltages vs. ambient temperature (2)



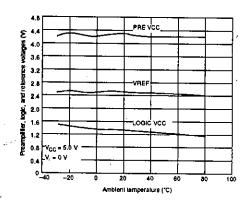
V_{REF} , preamplifier and logic reference voltages vs. ambient temperature (3)



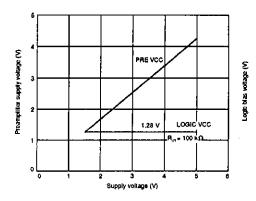
V_{REF} , preamplifier and logic reference voltages vs. ambient temperature (4)



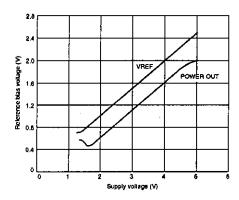
V_{REF}, preamplifier and logic reference voltages vs. ambient temperature (5)



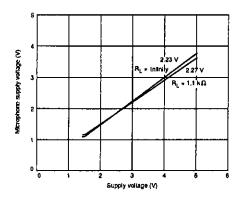
Preamplifier and logic reference voltages vs. supply voltage



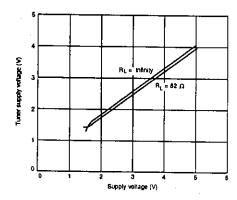
V_{REF} vs. supply voltage



Microphone supply voltage vs. supply voltage



Tuner supply voltage vs. supply voltage



FUNCTIONAL DESCRIPTION

Recording Preampliflers

The recording preamplifiers are coupled to an automatic gain control (AGC) circuit, which keeps the recording signal amplitude relatively constant for large variations in input amplitude. The AGC response time is set by a capacitor connected to AGC TIME. The gain and frequency responses of the recording preamplifiers are adjustable using the feedback networks connected to REC NF1 and REC NF2. The amplified recording signals are output on REC OUT1 and REC OUT2.

Microphone preamplifiers

MIC IN1 and MIC IN2 are the microphone preamplifier inputs. The gain and frequency responses are adjustable using the feedback networks connected to PRE NF1 and PRE NF2. The amplified microphone signals are output on MIC OUT1 and MIC OUT2 and are used as a signal source for the recording preamplifiers.

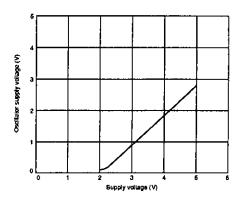
Tuner preamplifiers

TUNER IN1 and TUNER IN2 are the tuner preamplifier inputs. The buffered and amplified signals are output on MIC OUT1 and MIC OUT2 and are used as a signal source for the recording preamplifiers.

Playback Preamplifiers

PRE IN FWD L CH and PRE IN FWD R CH are the forward playback head preamplifier inputs, and PRE IN REV L CH and PRE IN REV R CH, the reverse playback head preamplifier inputs. The gain and frequency responses are adjustable using the feedback networks connected to PRE NF1 and PRE NF2. The amplified playback signals are output on PRE OUT1 and PRE OUT2.

Blas oscillator supply voltage vs. supply voltage



Power Amplifiers

Output level control

Audio signals are attenuated using an electronic variable resistor (EVR) before being amplified. The level of this attenuation is controlled by the voltage on EVR CONT. An input voltage range of 0 to 1.3 V corresponds to the attenuation range from maximum to minimum attenuation, respectively.

Headphone and speaker amplifiers

The power amplifier section comprises three separate amplification channels—two for driving stereo headphones and one common amplifier for driving a low-impedance speaker. Signals are input to the headphone amplifiers on POWER IN1 and POWER IN2 and output on POWER OUT1 and POWER OUT2. The gain and frequency responses of the headphone amplifiers are adjustable using the feedback networks connected to POWER NF1 and POWER NF2. POWER NF1 and POWER NF2 should be connected to VREF through coupling capacitors. The two headphone outputs should be mixed together, then input on SPKR DRIVE IN. The resulting amplified mono signal is output on SPKR DRIVE.

Reference Voltages

Power amplifier reference voltage

The reference voltage at VREF is biased to $0.5V_{CC}$ and can be used as a virtual earth because the output impedance is less than 1 Ω .

The capacitor on FILTER is used to reduce ripple on the reference voltage. A value between 22 and 33 μF is recommended. Larger values increase the power supply startup time. Smaller values can cause an unacceptable amount of ripple.

External microphone supply

The reference voltage at MIC VCC is biased to $0.75 V_{\rm CC}$ and is used to power external electret condenser microphones.

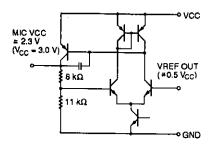


Figure 1. Microphone supply

Blas oscillator supply

The reference voltage at OSC VCC is biased to $(V_{\rm CC}-2)$ V and is used to power an external recording bias oscillator.

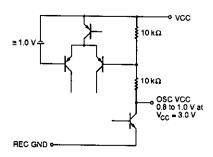


Figure 2. Oscillator supply

Operating Mode Inputs

Tape/tuner/standby modes

VCC/TAPE/TUNER controls the primary operating mode of the LA4583M. When VCC/TAPE/TUNER is connected directly to VCC, the tuner circuits are selected and the tuner supply voltage is output on TUN VCC OUT. TUN VCC CONT can also be used with an external transistor, such as an 2SA608, to supply other circuits while in tuner mode. When VCC/TAPE/TUNER is connected to VCC through an 82 k Ω resistor, the tape circuits are selected. When this pin is left unconnected, the device goes into standby mode.

Microphone recording monitor

REC MONITOR ON OFF enables and disables monitoring of the microphone signal during recording. When REC MONITOR ON OFF is connected to ground, the signal being recorded is sent to the headphone outputs.

Preamplifier reference

The reference voltage at PRE VCC is biased to 0.42V_{CC} and is used to power the internal playback, microphone and tuner preamplifiers.

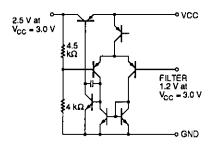


Figure 3. Preamplifier reference

Playback direction and recording control

When REC/PLAY/REV CONT is left unconnected, forward playback is selected. When REC/PLAY/REV is connected to ground through a 20 k Ω resistor, reverse playback is selected. When REC/PLAY/REV is connected to VREF, recording mode is selected.

Mute

Muting is activated when the operating mode is changed, greatly reducing mode switching noise. Muting is released after a delay determined by the capacitor on MUTE TIME C. The delay is approximately 0.4~s with a $4.7~\mu F$ capacitor. The voltage change on this pin and the resulting mute duration is shown in figure 4.

DESIGN INFORMATION

Voltage Gain in Different Modes Playback preamplifiers and power amplifiers

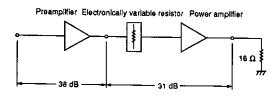


Figure 5. Preamplifier gain

The preamplifier voltage gain can be adjusted using a resistor on the negative feedback inputs. The voltage gain should be kept above 25 dB, otherwise spurious oscillation can occur. The power amplifier gain is fixed at 31 dB. Reducing the power amplifier load to 32 Ω increases the gain by approximately 1 dB.

Common power amplifier

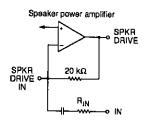


Figure 6. Power amplifier gain

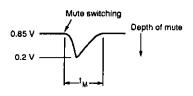


Figure 4. Mute timing

Voltage gain calculation

The gain of the common power amplifier can be varied by altering the feedback resistor R_{IN} as shown in the following equation. R_{IN} can be set to ∞ if a low output impedance is required.

 $VG = 20 \times \log_{10}((20 \text{ k}\Omega) / R_{\text{IN}})$

Microphone and recording amplifiers

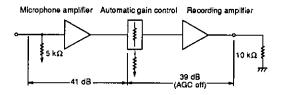


Figure 7. Microphone amplifier gain

The input impedance of the microphone amplifier is $5 \text{ k}\Omega$. As the feedback resistor is $27 \text{ k}\Omega$, care should be taken to limit the gain when using electret condenser microphones.

Voltage gain calculation

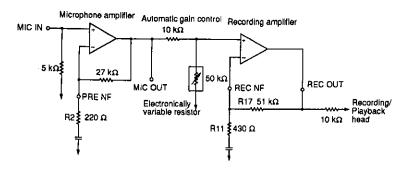


Figure 8. Microphone amplifier

Microphone amplifier gain

 $VG_{MR} = 20 \times log_{10}((27 k\Omega) / R22)$

The microphone amplifier gain should be above 25 dB.

Recording amplifier gain

 $VG_{RR} = 20 \times log_{10}((51 \text{ k}\Omega) / \text{R11})$

Tuner and recording amplifiers

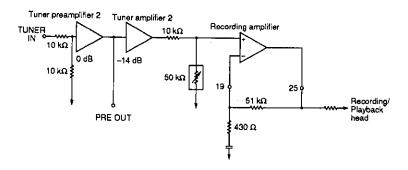


Figure 9. Tuner amplifier

The total gain of the tuner preamplifier and recording amplifier can be calculated as follows.

 $VG_{TR} = (20 \times log_{10}((51 \text{ k}\Omega) / R11) - 14 \text{ dB}$

APPLICATION NOTES

Typical Application

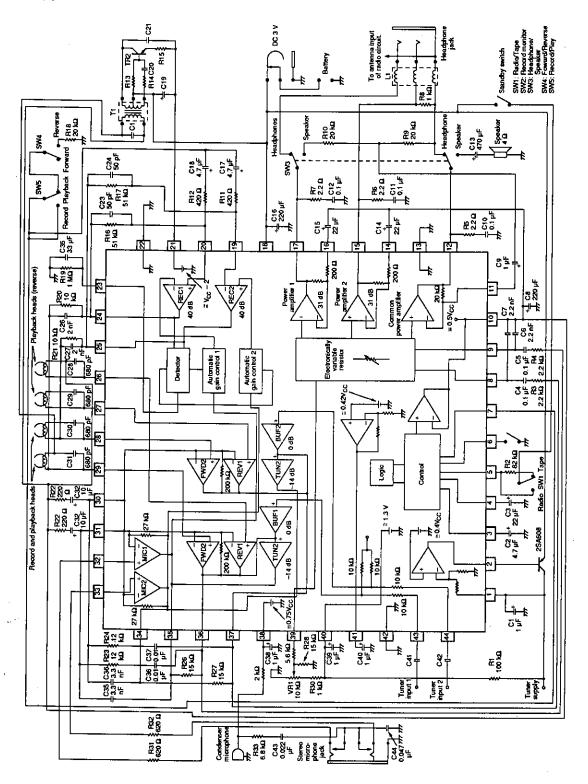


Figure 10. Typical application

Description of external components

Microphone preampliflers

MIC VCC supplies power to an external electret condenser microphone through a current limiting resistor and a decoupling capacitor, C38. A value between 1 and 3.3 μ F should be chosen. R33 and C43 couple the audio signal to the MIC IN inputs and should be selected according to the microphone used. R31 and R32 are included as static discharge protection for the preamplifier inputs. A value between 510 and 1 k Ω should be used. R22, R23, C32 and C33 are placed in the negative feedback path of the microphone and playback preamplifiers to create a low-frequency cutoff of 40 Hz. Values should be chosen from those in the following table.

R22, R23	C32, C33
100 Ω	3 3 μF
200 Ω	22 μF
300 Ω	10 μF
400 Ω	4.7 μF

C35 and C36 are highpass filter capacitors which, in conjunction with the internal 27 k Ω feedback resistors, remove noise from the microphone inputs. A value between 2.2 and 4.7 μF should be used. R22 and R23 set the gain of the microphone and playback preamplifiers. A value between 100 and 390 Ω should be used. R26 and R27 are used to connect the microphone outputs to the power amplifier inputs for record mode monitoring. A value of 15 k Ω is recommended.

Playback preamplifiers

C28 to C31 are the playback head highpass filter capacitors and should be selected to suit the head used. Values between 330 and 2200 pF are recommended. C36 and C37 are the equalization curve setting capacitors. A value should be selected to suit the manufacturer's specifications.

Recording amplifiers

R11, R12, R16 and R17 set the recording amplifier voltage gain. Values of 420 Ω for R11 and R12, and 51 k Ω for R16 and R17 are recommended. C23 and C24 are placed in parallel with R16 and R17 to reduce ringing on the recording amplifier outputs. A value between 30 and 50 pF should be used. C17 and C18 are negative feedback capacitors and should be connected so that the positive poles are connected to pins 19 and 20. A value of 4.7 μ F is recommended. C26 and C27 perform highpass filtering on the outputs and should be selected to suit the manufacturers specifications. R20 and R21 represent the load on the outputs. A value of 10 k Ω is recommended.

The AGC circuit coupled to the recording amplifier has a response that is dependent on the values of R19 and C25. R19 sets the attack time and C25 sets the recovery time. A value between 300 k Ω and 2 M Ω should be used for R19, and 22 to 47 μ F, for C25.

Power amplifiers

C4 and C5 couple the preamplifier outputs to the power amplifier inputs. Connect the positive poles towards the preamplifier inputs and select a value between 0.1 and 1 μF. R3, R4, C6 and C7 form highpass filters for audio signals entering the power amplifiers. A value between 0 and 3 $k\Omega$ should be used for R3 and R4, and 560 to 2200 pF, for C6 and C7. C14 and C15 are the negative feedback capacitors for the headphone amplifiers. Connect the positive poles toward the outputs and choose a value between 10 and 22 µF. The lowfrequency cutoff is approximately 36 Hz with 22 µF capacitors. On the output of each channel is an RC network designed to suppress unwanted oscillation. A value between 1.5 and 3.0 k Ω should be chosen for resistors R5 to R7. The corresponding capacitors, C10 to C12, should be 0.1 µF ceramic capacitors.

The two output channels of the headphone amplifiers are combined and input on SPKR DRIVE IN, the input to the common power amplifier. The recommended circuit is shown in figure 11. Resistor values should be selected from the following equations.

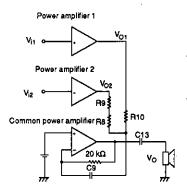


Figure 11. Speaker drive circuit

Note $R8 = 1 \text{ to } 2 \text{ k}\Omega$ $R9 \text{ and } R10 = 20 \text{ k}\Omega$ $V_o = 31 \text{ dB} \times ((40 \text{ k}\Omega) / R9) \times V_i$ If R9 = R10 and R9 >> R8, then $V_o = (V_{o1} / R10 + V_{o2} / R9) \times (20 \text{ k}\Omega)$ If $V_{o1} = V_{o2} = V_o$, then $V_{i1} = V_{i2} = V_i$ $V_o = V_o$, $(2/R9) \times (20 \text{ k}\Omega)$ $= V_o$, $(4/R9) \times (20 \text{ k}\Omega)$

The output of the common amplifier is used to drive a low-impedance speaker through a coupling capacitor, C13. A value between 220 and 470 μ F should be selected, however, a 470 μ F capacitor is recommended if high output is required.

Tuner preamplifiers

C41 and C42 are coupling capacitors for the TUN IN1 and TUN IN2 inputs. A value between 0.01 and 0.1 μ F should be used.

Mode selection and switching

R2 is used with VCC/TAPE/TUNER CONT. This tristate input requires a resistor connected to VCC to activate the tape circuits. A value between 70 and 90 k Ω can be chosen, however, 82 k Ω is the recommended value.

R18 is used with REC/PLAY/REV CONT. This tristate input requires a resistor connected to ground to select reverse playback. A value between 15 and 24 k Ω can be chosen, however, 20 k Ω is the recommended value.

R1 is the tuner power supply startup resistor. If not present, tuner power supply operation is not guaranteed. A value between 50 and 100 $k\Omega$ should be used.

The capacitor on MUTE TIME C, C2, determines the mute time. A value of 4.7 μF will result in a mute time of approximately 0.4 s.

External bias oscillator circuit

The values of all components related to the external bias oscillator circuit, R13 to R15 and C19 to C22, should be selected according to the manufacturers specifications.

Electronic variable resistor (EVR)

R28 sets the minimum output signal level. A value of 15 k Ω is recommended. R29 sets the maximum output signal level for which a value of 5.6 k Ω is recommended. R30 has been added to protect the EVR input from damage due to static discharge.

Supply and reference voltages

C1 is the tuner power supply decoupling capacitor. A value between 1 and 10 μF should be used. C3 is the ripple filter capacitor for the reference voltage for which a value of 22 μF is recommended. C8 is the reference bias decoupling capacitor for which a value of 220 μF is recommended. C16 decouples the LA4583M supply. Again, a value of 220 μF is recommended.

C39 is both the logic supply decoupling capacitor and oscillation suppressor for which a value between 0.01 and 0.1 μF should be used. C40 decouples the preamplifier bias supply for which a value between 4.7 and 10 μF should be selected.

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